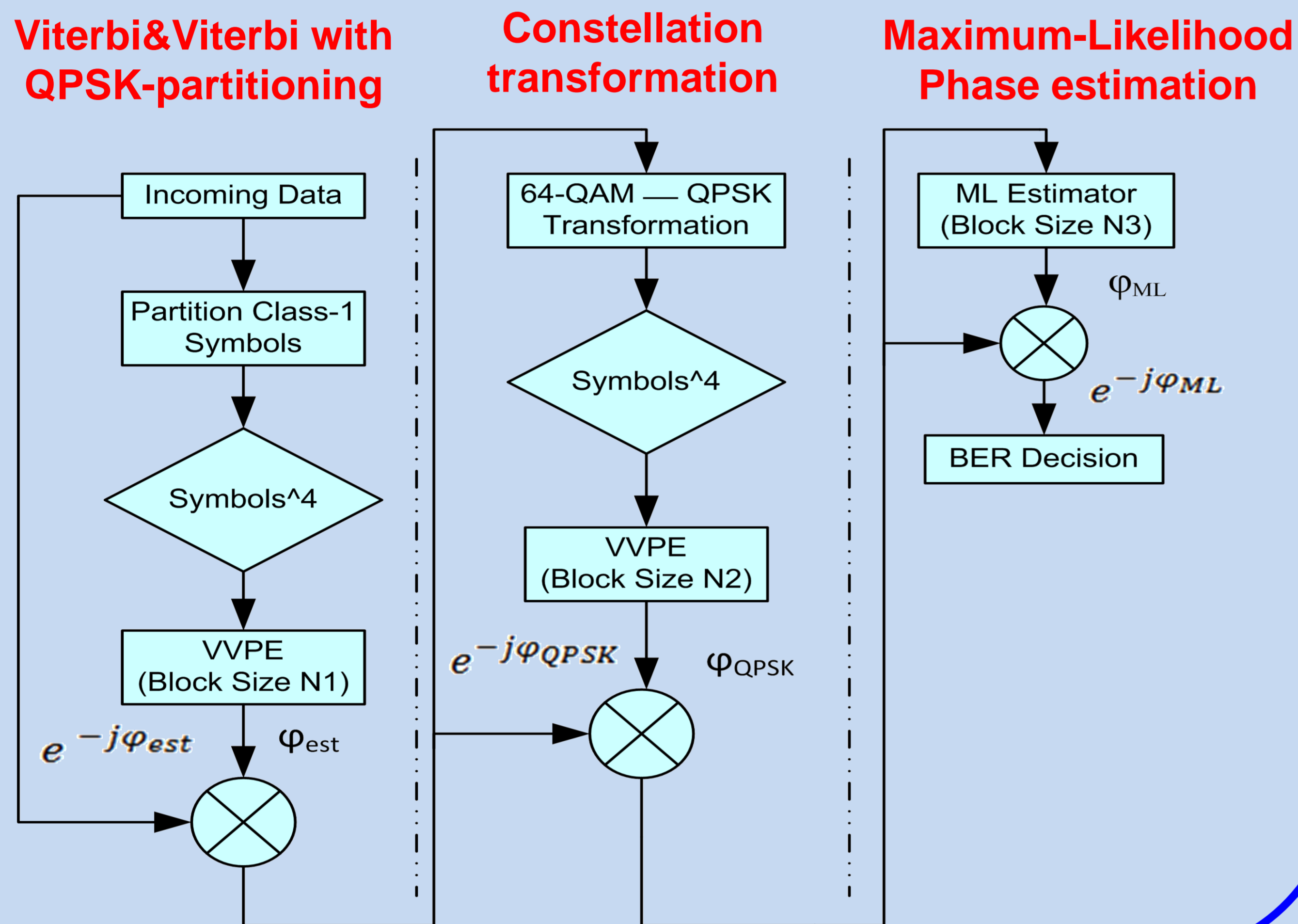
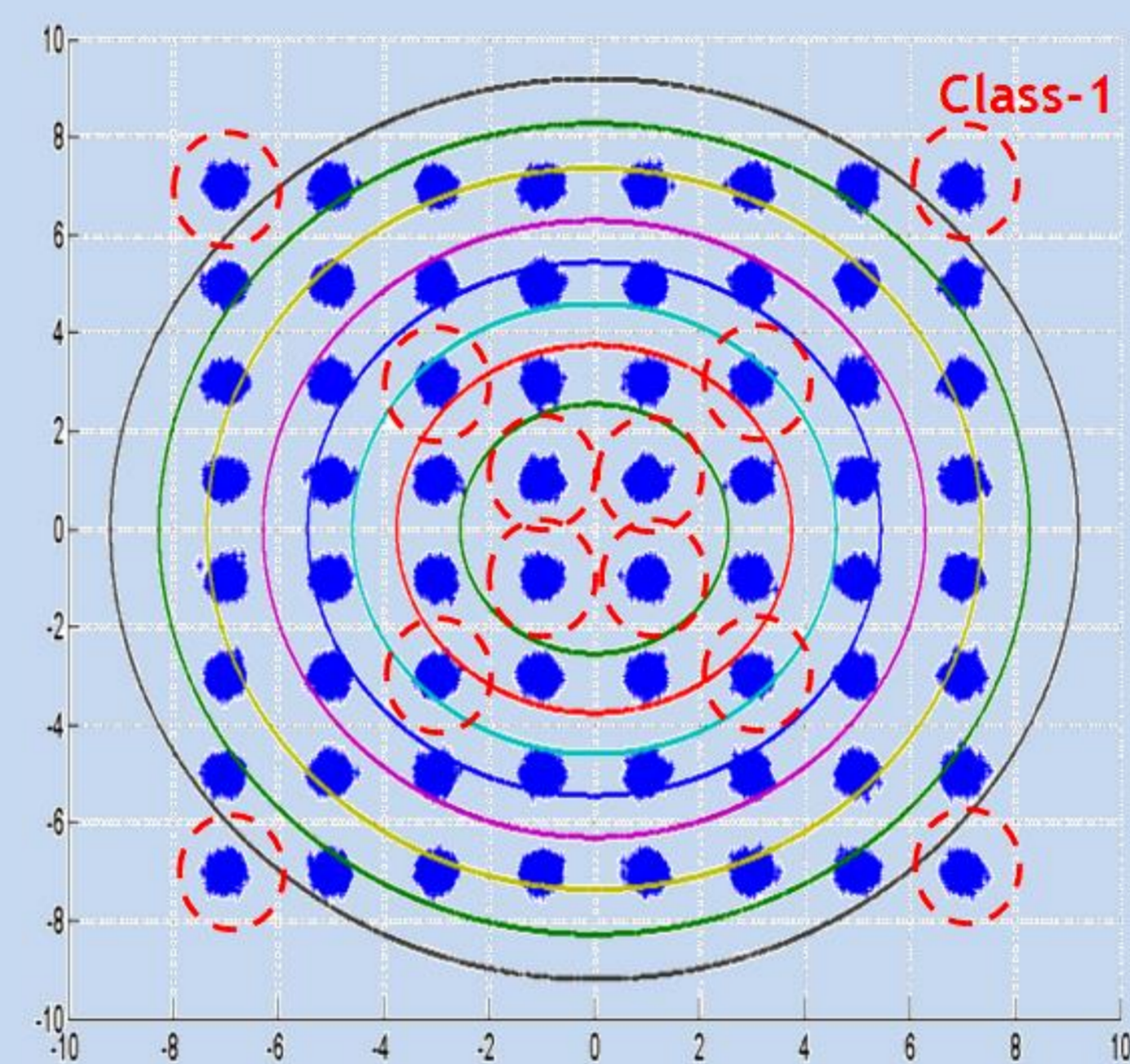


**Abstract** - A novel three-stage digital feed-forward carrier recovery algorithm based on the transformation of 64-QAM constellation into QPSK is proposed. For 1 dB penalty at BER=10<sup>-2</sup>, it can tolerate a linewidth-times-symbol-rate product of 4.5·10<sup>-5</sup>, making it possible to operate 32-Gbaud optical 64-QAM systems with current commercial tunable lasers.

## Block diagram of the algorithm



## QPSK partitioning and MLE stages



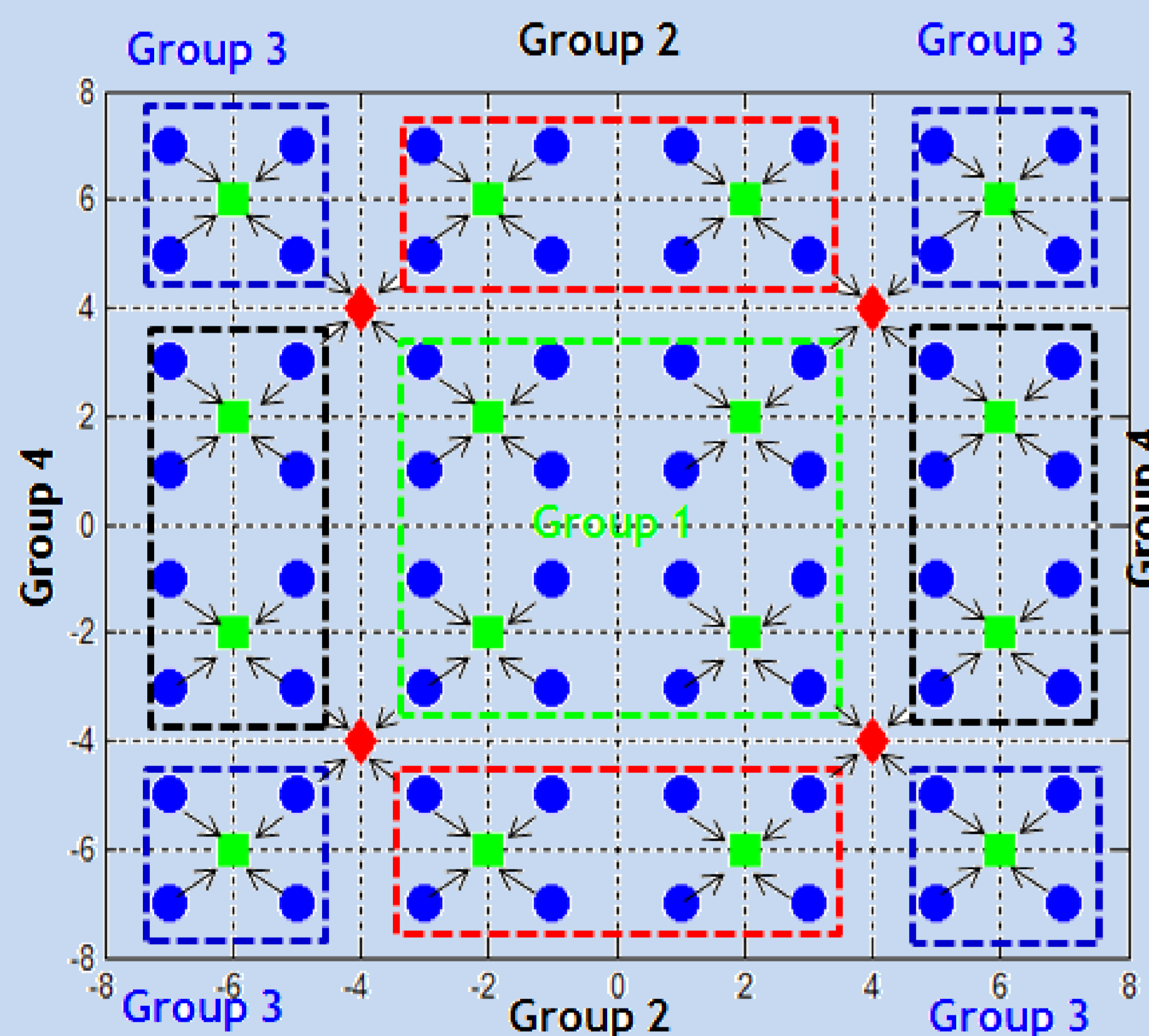
- Class-1 symbols have modulation angles equal to  $\pi/4 + m \cdot \pi/2$  ( $m = 0 \dots 3$ ).
- 12 out of the 16 symbols lying at the vertices of squares are used in the Viterbi&Viterbi algorithm:

$$\varphi_{est, class1} = \frac{1}{4} \arg \sum_{k=1}^{N_1} \frac{X_k^4}{|X_k^4|}$$

- Performance of the estimators can be further improved by adding an MLE stage:

$$\varphi_{ML} = \tan^{-1} \left( \frac{\text{imag}(z)}{\text{real}(z)} \right) \quad \text{with} \quad z = \frac{1}{4} \arg \sum_{k=1}^{N_2} x_k \cdot y_k$$

## 64-QAM → QPSK Constellation Transformation



From 64-QAM to 16-QAM:

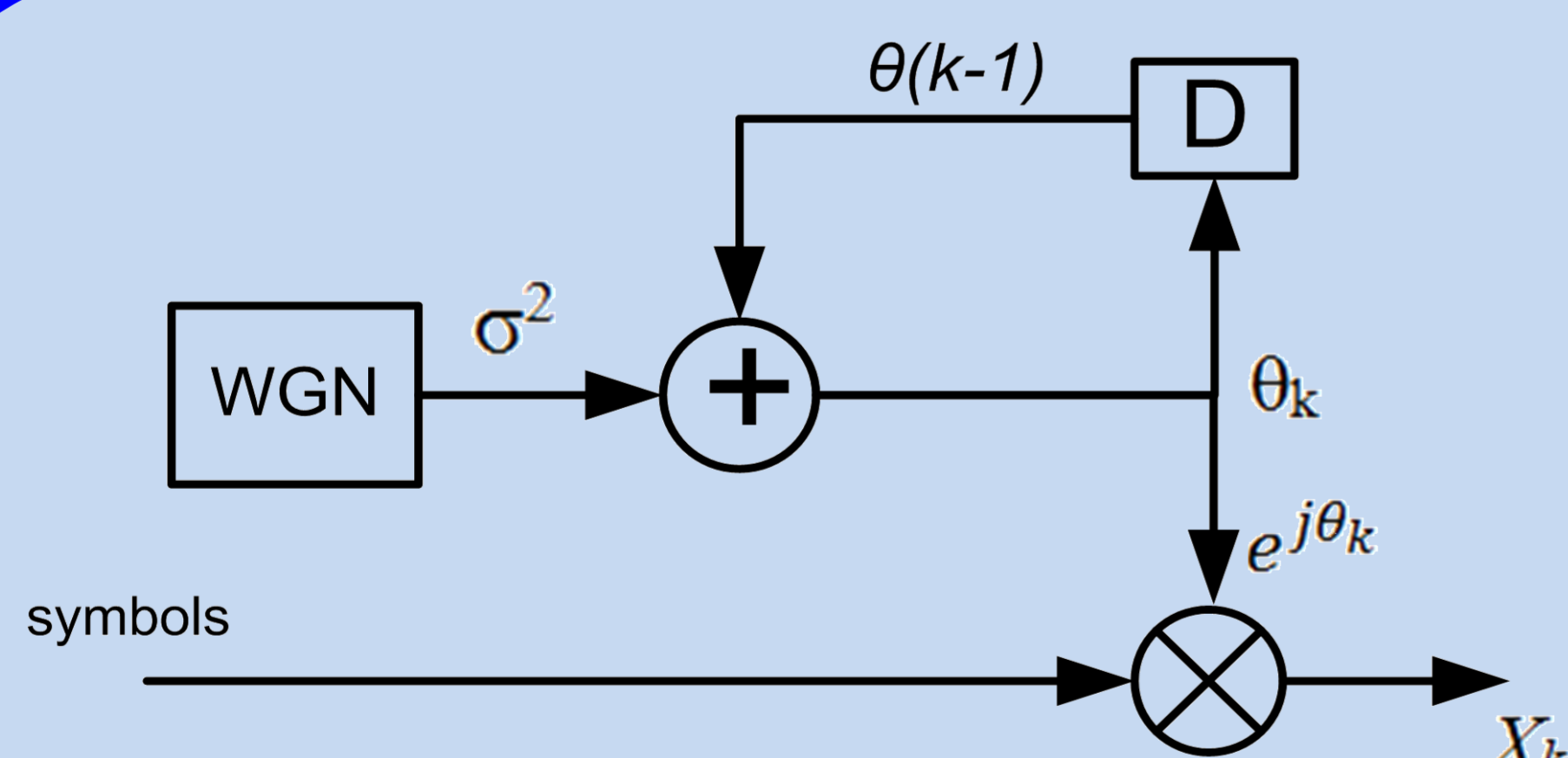
$$Y = Y_{1r} - \text{sgn}(Y_{1r} - 2 \text{sgn}(Y_{1r})) + j(Y_{1i} - \text{sgn}(Y_{1i} - 2 \text{sgn}(Y_{1i}))) + Y_{2r} - \text{sgn}(Y_{2r} - 2 \text{sgn}(Y_{2r})) + j(Y_{2i} - \text{sgn}(Y_{2i} - 6 \text{sgn}(Y_{2i}))) + Y_{3r} - \text{sgn}(Y_{3r} - 6 \text{sgn}(Y_{3r})) + j(Y_{3i} - \text{sgn}(Y_{3i} - 6 \text{sgn}(Y_{3i}))) + Y_{4r} - \text{sgn}(Y_{4r} - 6 \text{sgn}(Y_{4r})) + j(Y_{4i} - \text{sgn}(Y_{4i} - 2 \text{sgn}(Y_{4i})))$$

From 16-QAM to QPSK:

$$Y = Y_r - \text{sgn}(Y_r - 2 \text{sgn}(Y_r)) + j(Y_i - \text{sgn}(Y_i - 2 \text{sgn}(Y_i)))$$

- The constellation transformations are applied after frequency offset compensation between the LO and transmitter laser and after an initial phase noise correction using a coarse estimate (achieved using the Viterbi&Viterbi algorithm based on QPSK partitioning)

## Simulation Setup and results

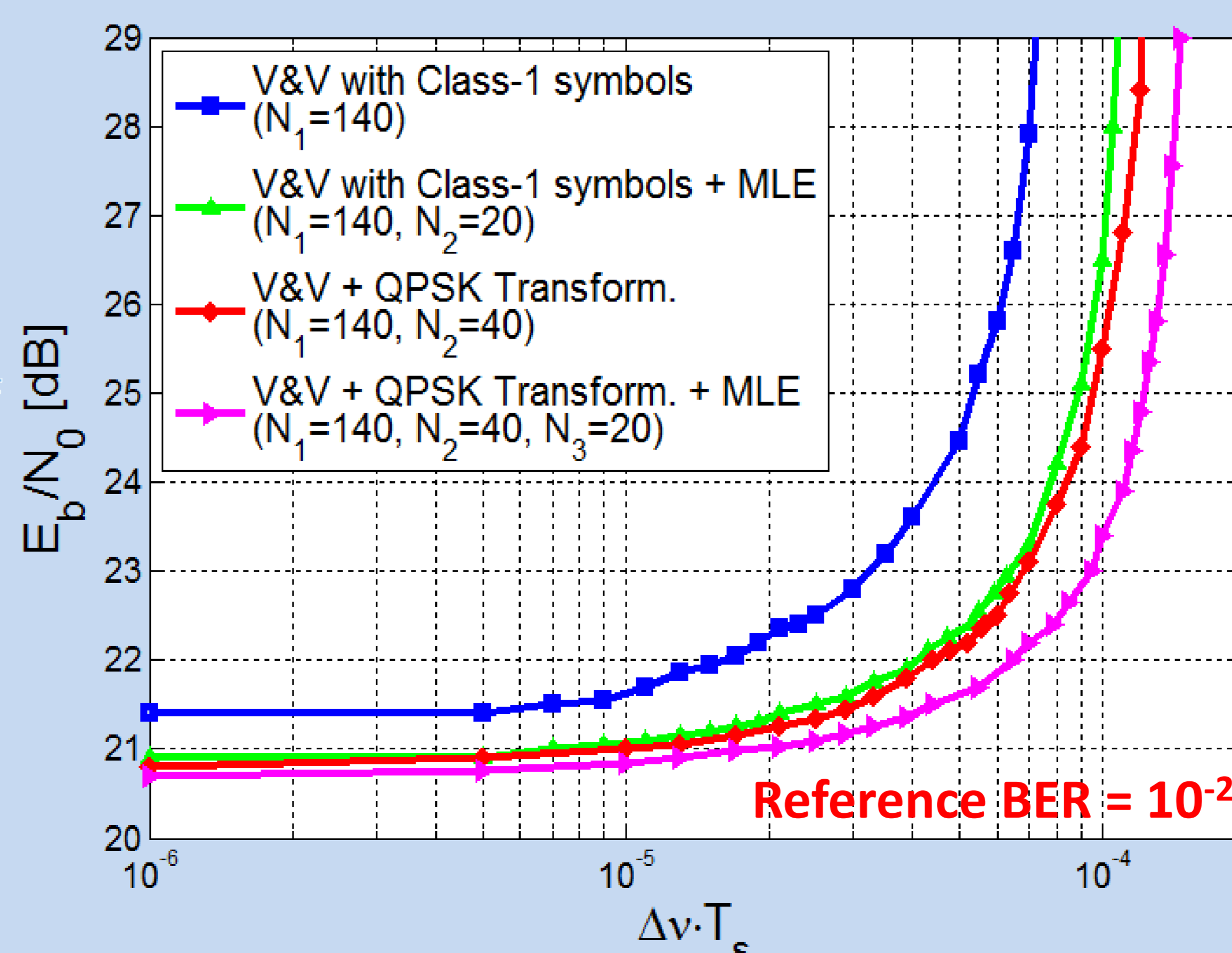


- $x_k$  is the data symbol that belongs to the set:  $(\pm a \pm j.b)$ ,  $a, b \in \{1, 3, 5, 7\}$ .
- $\eta_k$  is the AWG noise.
- $\theta_k$  is the laser phase noise, modeled as a Wiener process.
- $\Delta\nu$  is the combined laser linewidth of Tx laser and local oscillator.
- $T_s$  is the symbol period.

$$y_k = x_k e^{j\theta_k} + \eta_k$$

$$\theta_k = \sum_{i=-\infty}^k v_i$$

$$\sigma_f^2 = 2\pi\Delta\nu T_s$$



This work was supported by CISCO Systems within a SRA contract.

**Conclusion:** A novel low complexity algorithm for carrier phase estimation of 64-QAM has been presented and its performance analyzed through numerical simulations.

A linewidth times symbol duration product ( $\Delta\nu \cdot T_s$ ) equal to 4.5·10<sup>-5</sup> is tolerated for 1-dB penalty at BER equal to 10<sup>-2</sup>.

Assuming the industry-standard symbol rate of 32 Gbaud, this means that a total combined linewidth of over 1.3 MHz could be tolerated, making it possible to operate optical 64-QAM systems with current commercial tunable lasers.