

Coherent polarization-multiplexed formats: receiver requirements and mitigation of fiber non-linear effects

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- Multi-level modulation with coherent detection are clearly becoming the enabling solution for future high capacity long-haul system
- A key component in such systems is the ADC: relaxing the requirements will be beneficial reducing costs, complexity and power consumption
- Fiber non-linear effects are limiting the maximum reach but coherent detection offers new opportunities for mitigating the impact

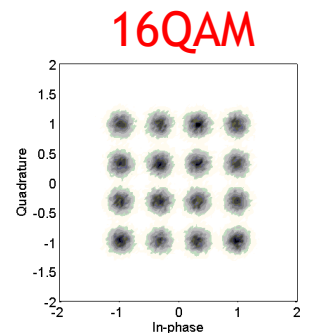
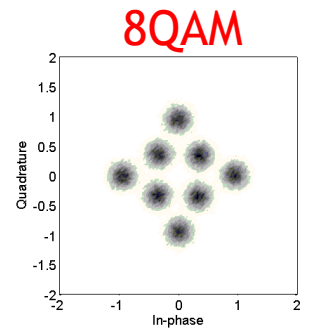
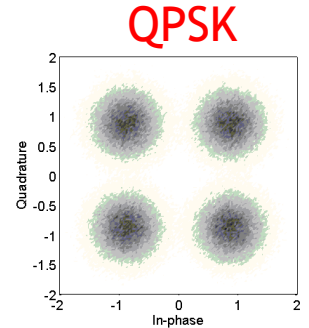


1. TX and RX set-up
2. Receiver requirements
3. Mitigation of fiber non-linear effects
4. Conclusions

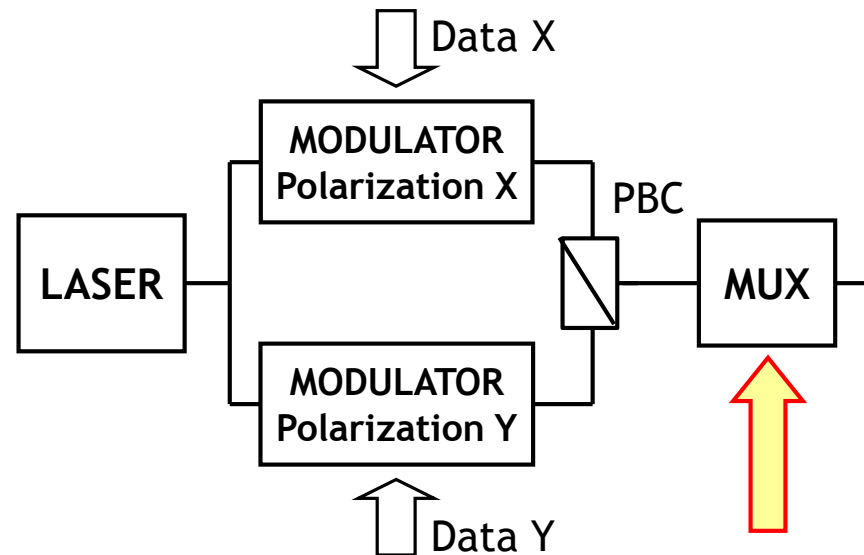


1. TX and RX set-up

- ▶ Symbol rate
 - ▶ 25 Gbaud + 11% overhead: 27.75 Gbaud
- ▶ Three modulation formats:
 - ▶ PM-QPSK → 111.0 Gbps (100 Gbps)
 - ▶ PM-8QAM → 166.5 Gbps (150 Gbps)
 - ▶ PM-16QAM → 222.0 Gbps (200 Gbps)
- ▶ Channel spacing
 - ▶ $\Delta f = 50$ GHz
- ▶ **Simulation details**
 - ▶ Independent PRBSs for each tributary (degree 16)
 - ▶ BER evaluation based on error counting
 - ▶ Reference BER = $4 \cdot 10^{-3}$
 - ▶ Fiber propagation: full band split-step method

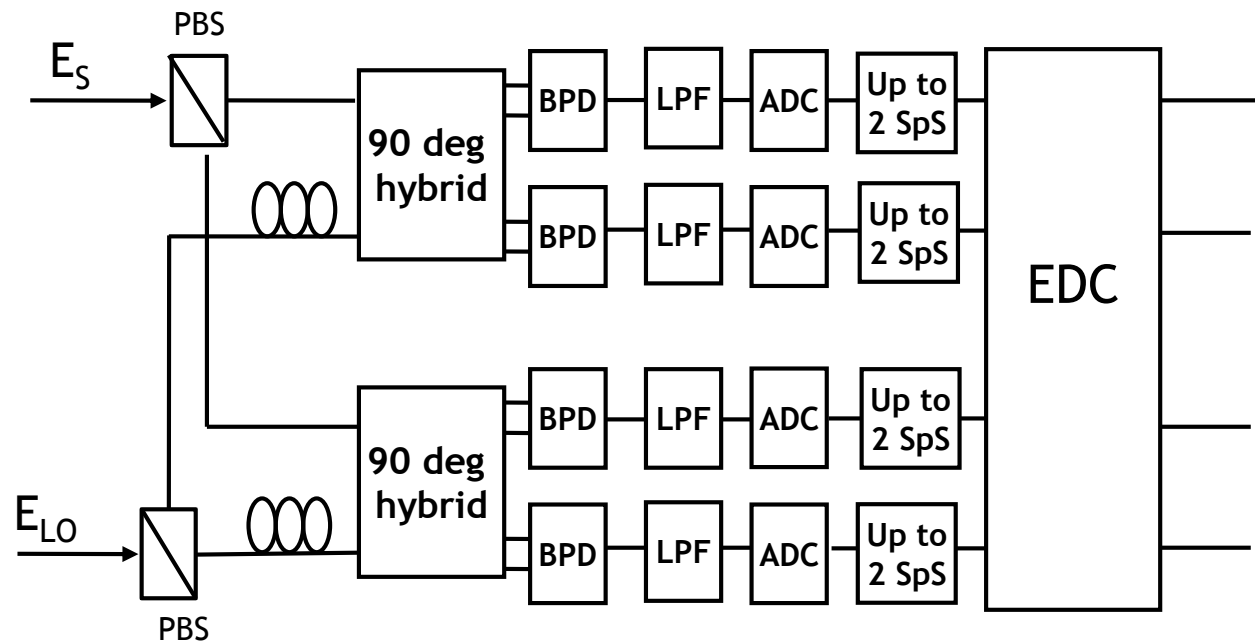


- ▶ Based on Nested Mach-Zehnder for QPSK and 16QAM
 - ▶ Binary driving for QPSK
 - ▶ Four level driving with pre-distortion for 16QAM
- ▶ Two cascaded Nested Mach-Zehnders for 8QAM



Optical Bandpass Filter
4th order Supergaussian

- ▶ Common opto-electronic front-end for all formats
- ▶ When ADC speed is reduced below 2 SpS, then up-sampling is performed to run DSP at 2 SpS
- ▶ Ideal clock recovery
- ▶ Ideal EDC





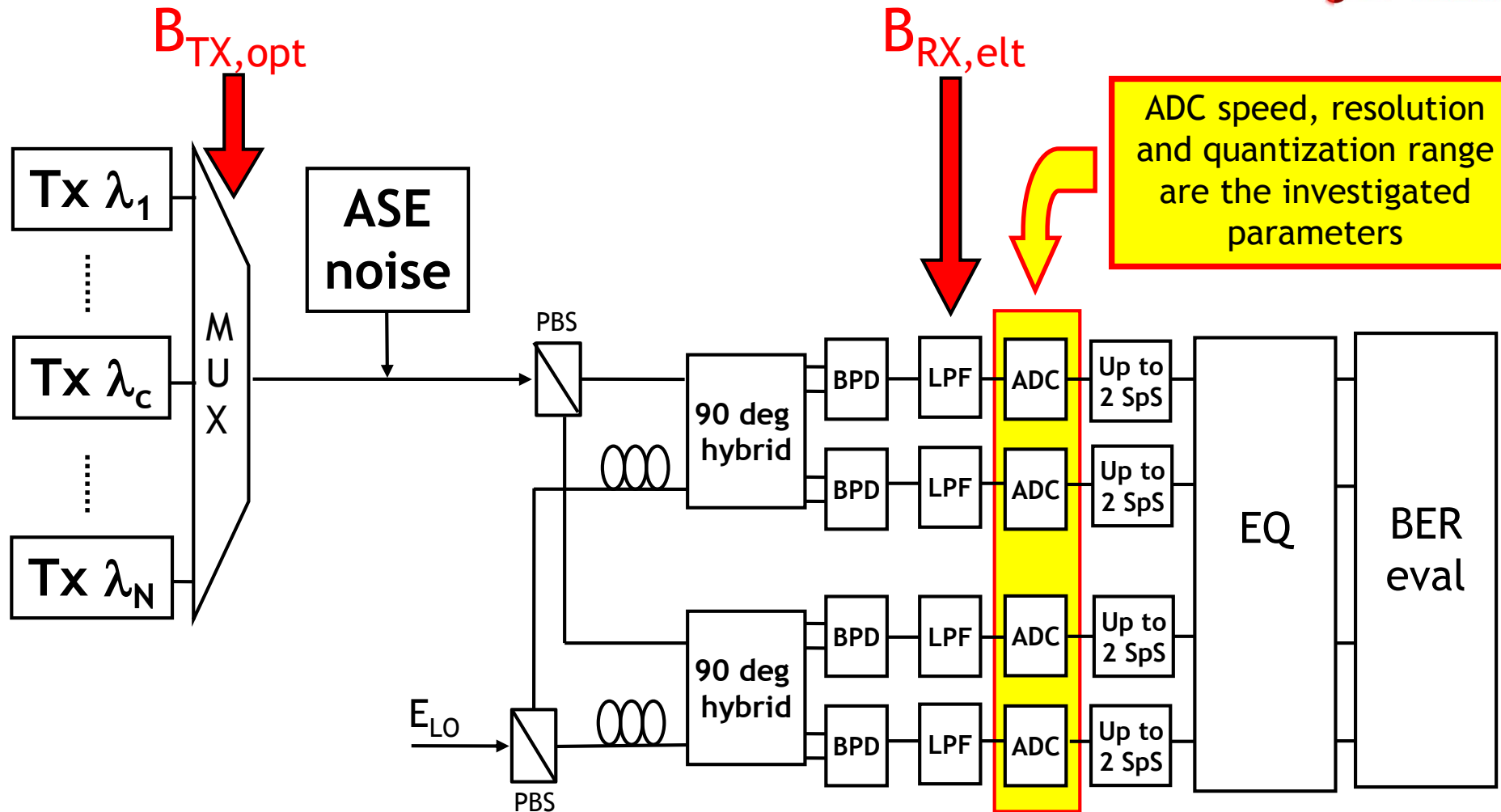
Equalizer updating and Symbol decision



- ▶ MIMO Equalizer
 - ▶ Ideal FIR with 15 taps
 - ▶ Updated through LMS
 - ▶ Training + Decision Driven
- ▶ Decision algorithms are specific for each format
 - ▶ PM-QPSK: single-threshold
 - ▶ PM-8QAM: maximum likelihood
 - ▶ PM-16QAM: multi-threshold

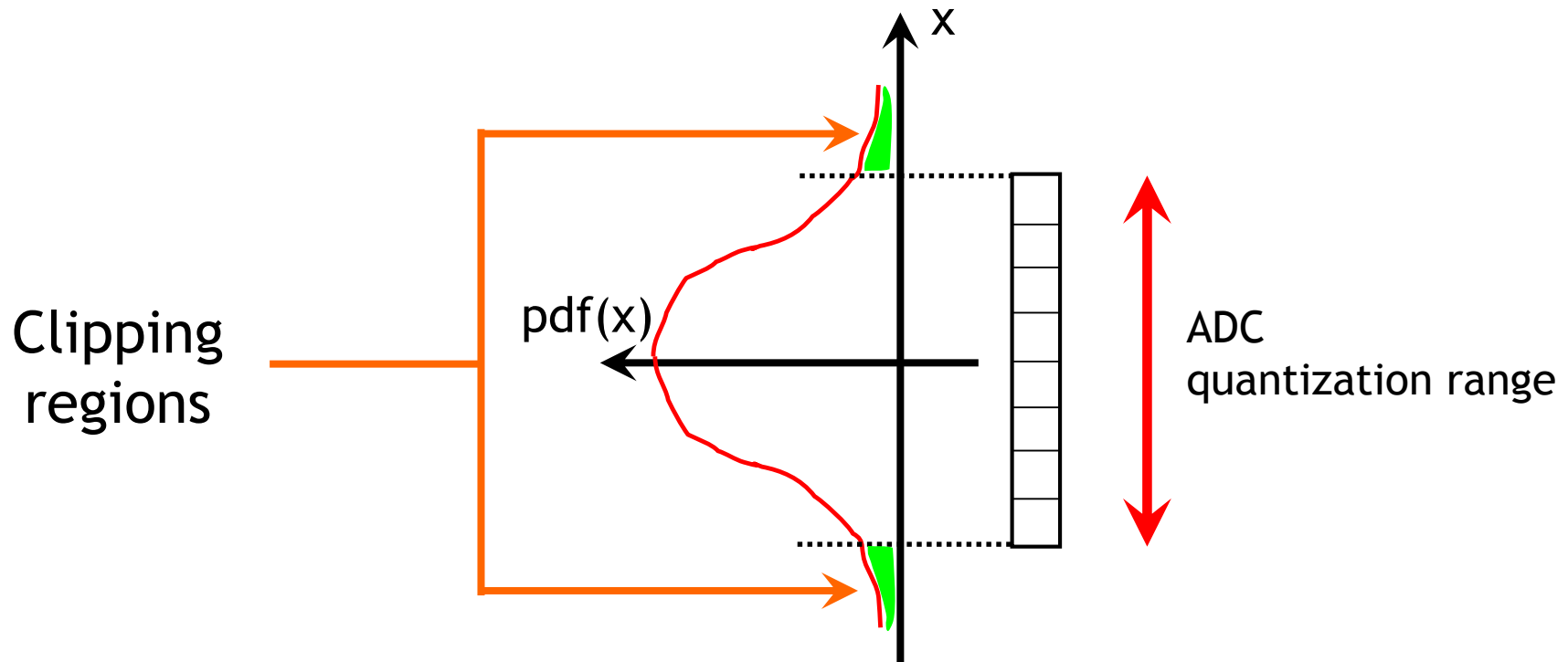


2. Receiver requirements



Optimal bandwidths of both Optical MUX and Electrical receiver filter are dependent on ADC parameters

- ▶ After propagation, signal components have a gaussian-like distribution (see [A] P4.07 at ECOC 2010)



We define as clipping percentage the portion of samples left out of the ADC quantization range

- ▶ We performed preliminary simulation in order to optimize the clipping percentage
- ▶ For all formats, in the range between 0.1% and 2%, performances does not change substantially at reference BER ($4 \cdot 10^{-3}$)

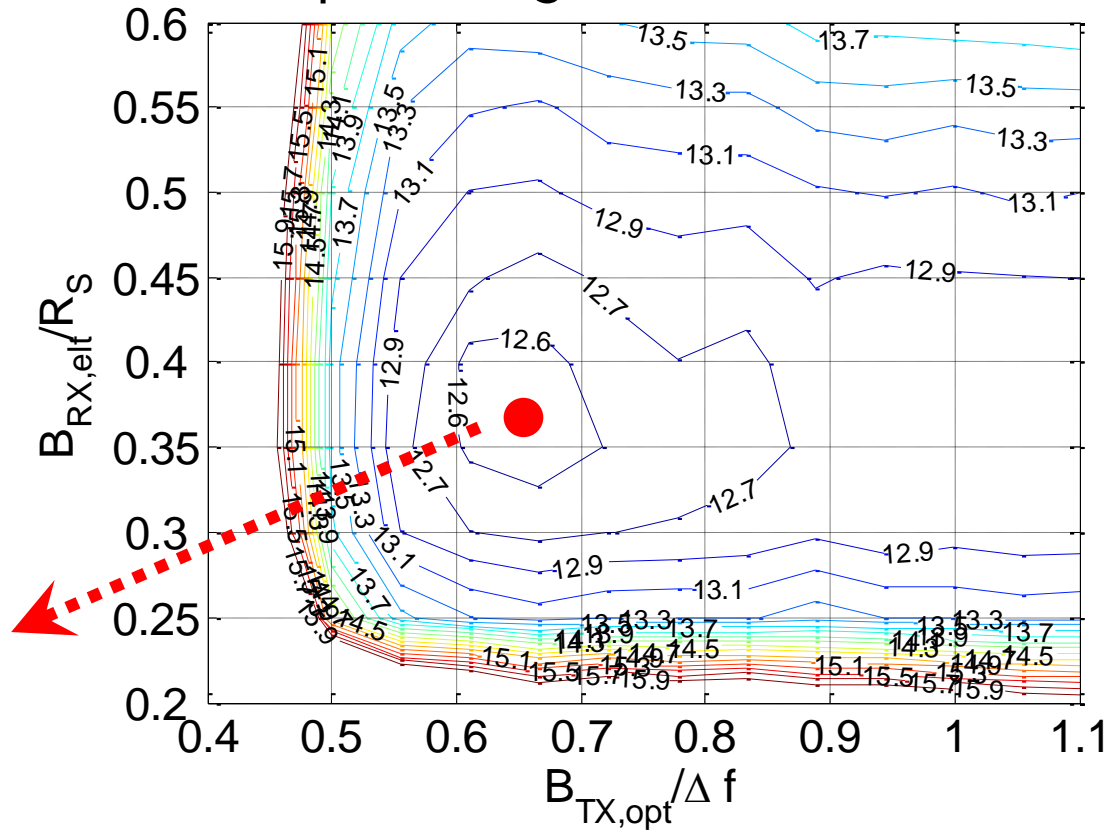
- ▶ For bandwidth optimization and non-linear analysis we have used the following clipping percentages:
 - ▶ 0.5% for PM-QPSK
 - ▶ 0.2% for both PM-8QAM and PM-16QAM

- ▶ For each modulation format
- ▶ ADC Speed \rightarrow SpS=[2.00 1.67 1.43 1.25 1.11]
- ▶ ADC Resolution \rightarrow $N_{\text{bit}}=[4 \ 5 \ 6 \ 7]$

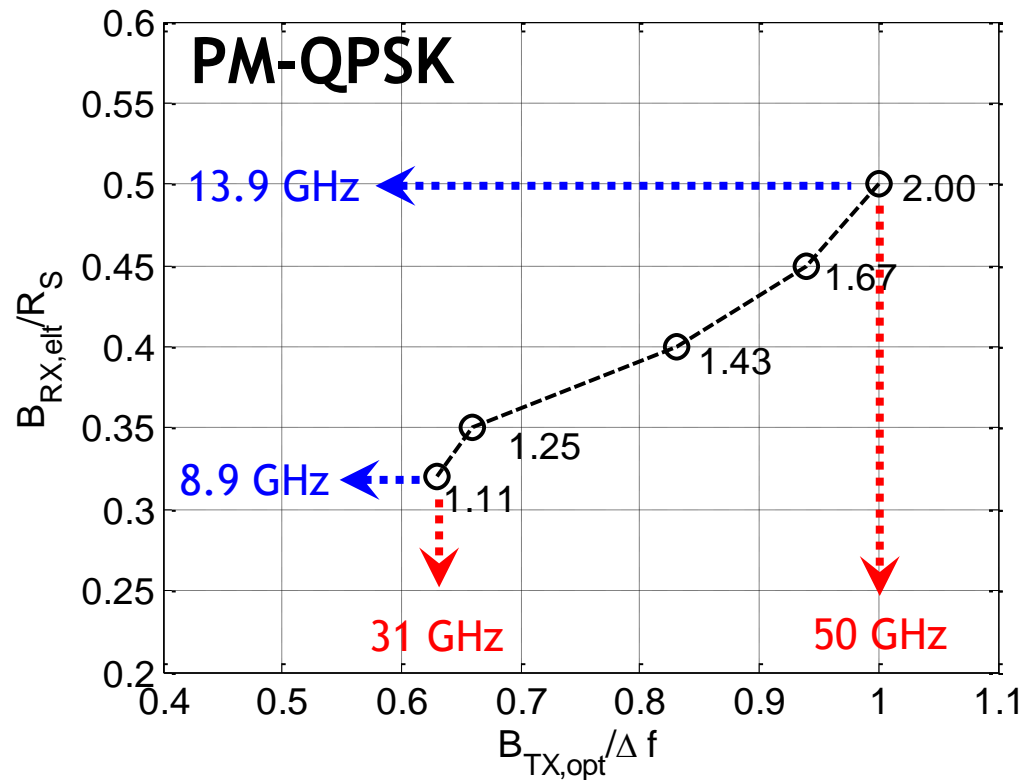
Contour plot of OSNR (in 0.1 nm) required to guarantee $\text{BER}=4 \cdot 10^{-3}$

PM-QPSK
SpS=1.25
 $N_{\text{bit}}=4$

$B_{\text{TX,opt}}=33 \text{ GHz}$
 $B_{\text{RX,elt}}=9.8 \text{ GHz}$

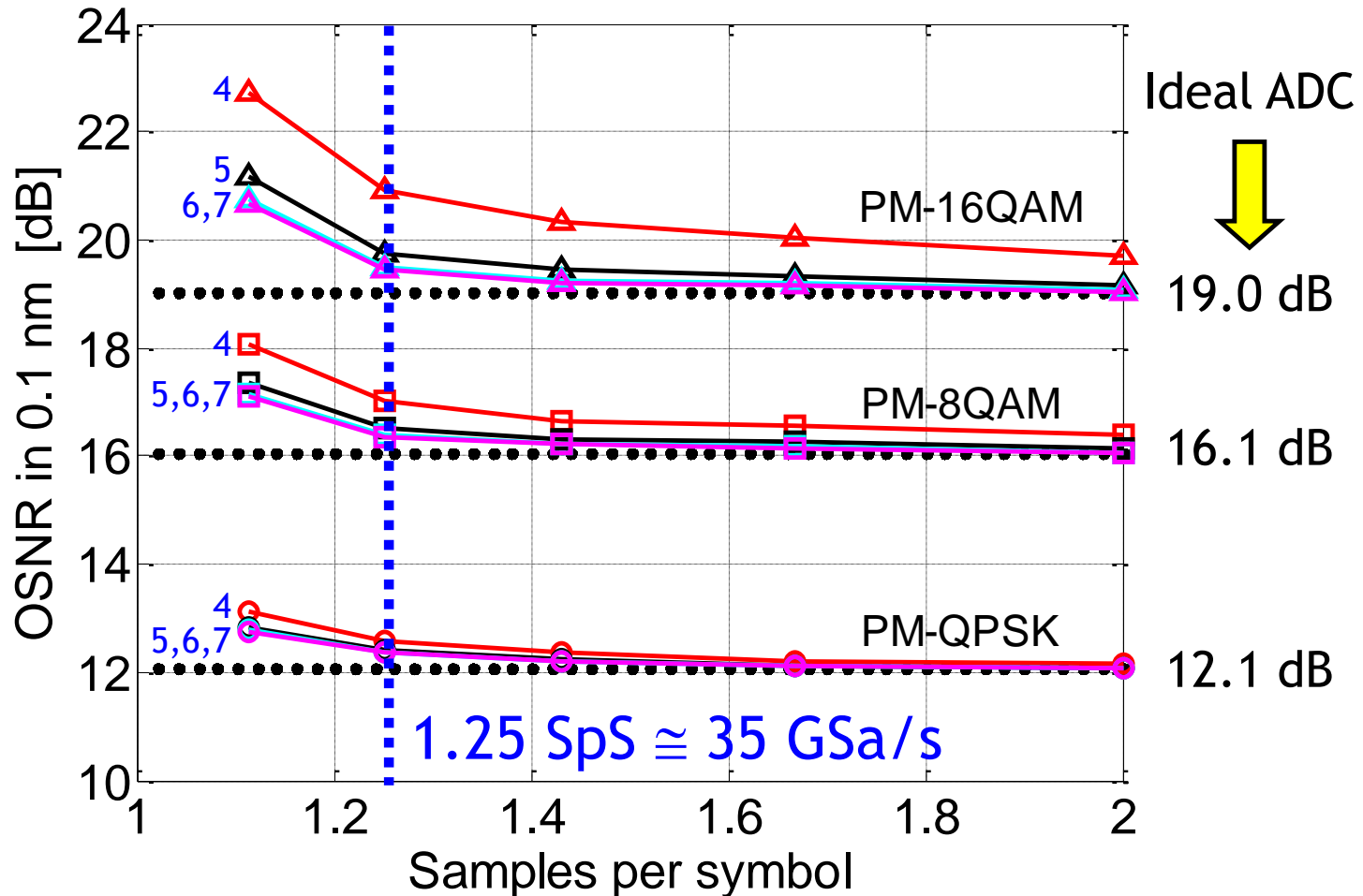


- ▶ Bandwidths optimization results did not show any dependence on ADC resolution

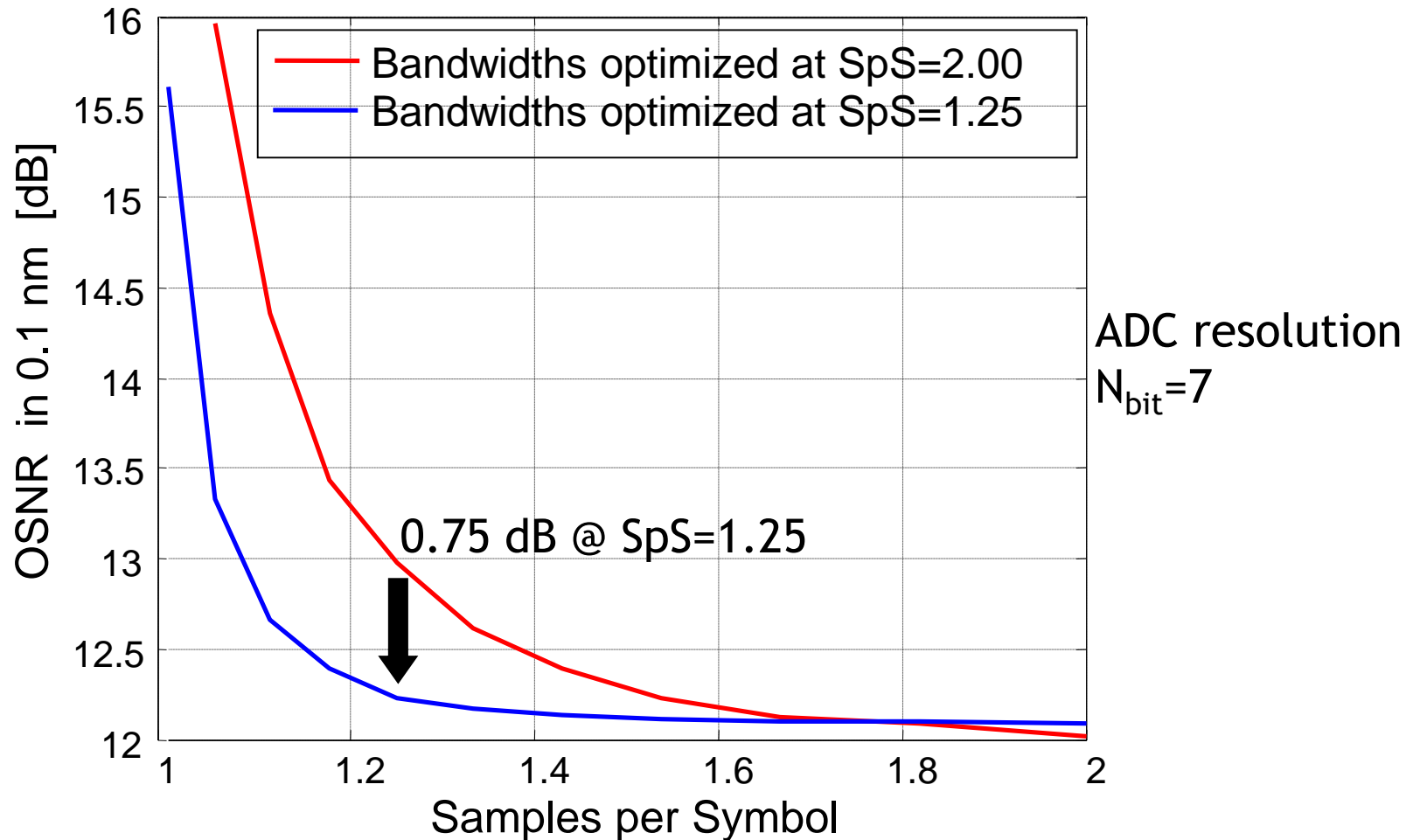


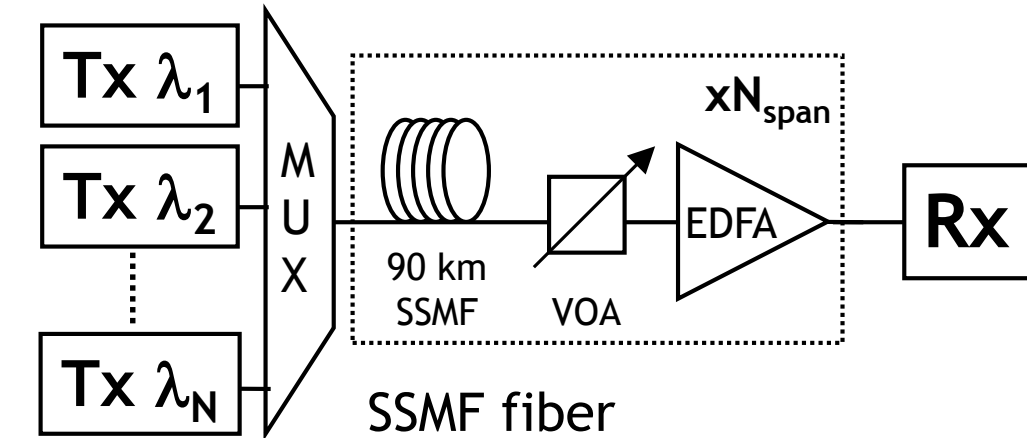
- ▶ Reducing ADC speed introduces aliasing in the system, that can be neutralized with tighter filtering

- ▶ We measured the OSNR (in 0.1 nm) required to guarantee $BER=4 \cdot 10^{-3}$



What if you do not properly optimize?





WDM
 $\Delta f = 50$ GHz

SSMF fiber

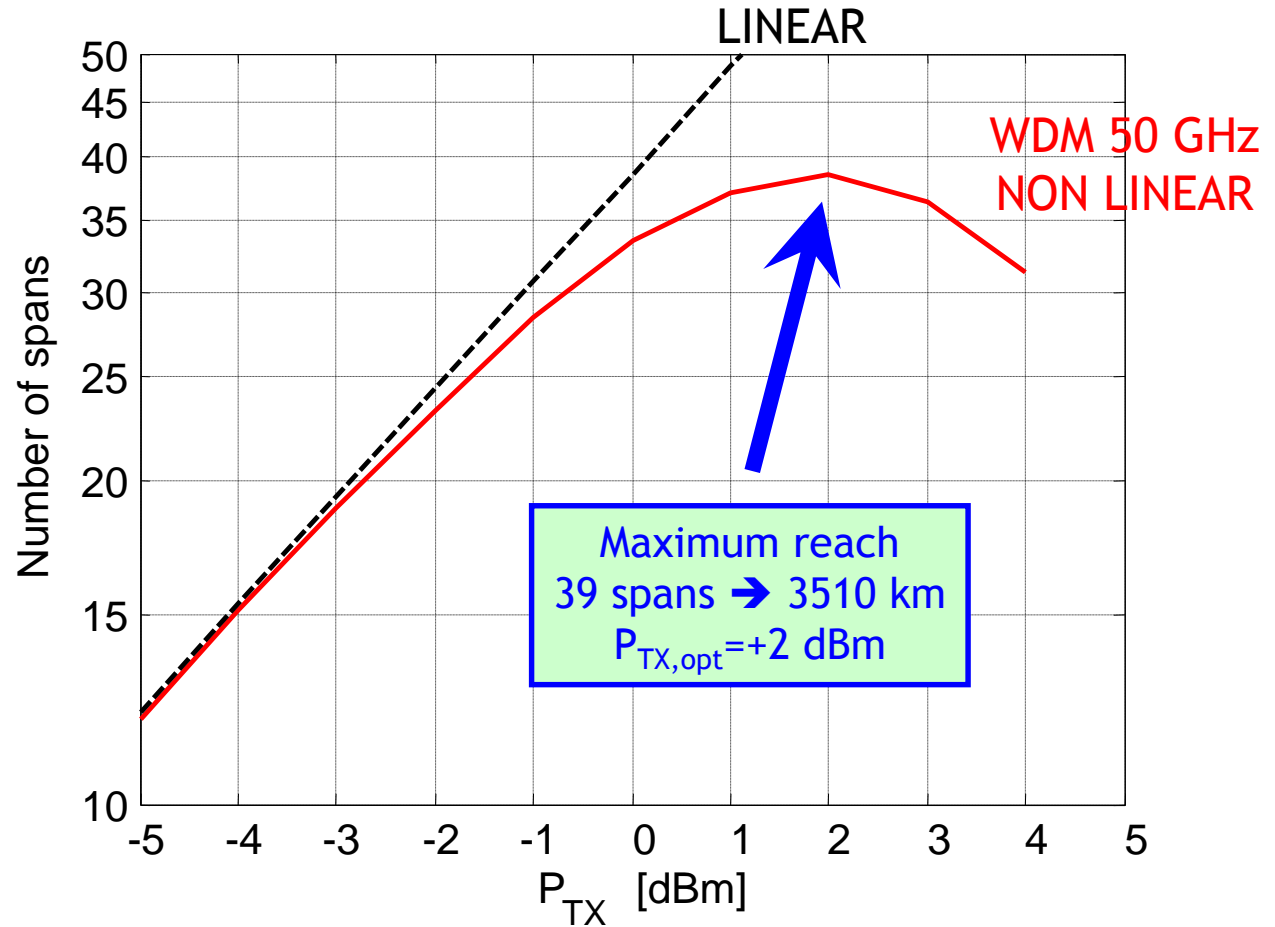
- $D = 16.7$ ps/nm/km
- $\alpha = 0.22$ dB/km
- $\gamma = 1.3$ 1/W/km

EDFA

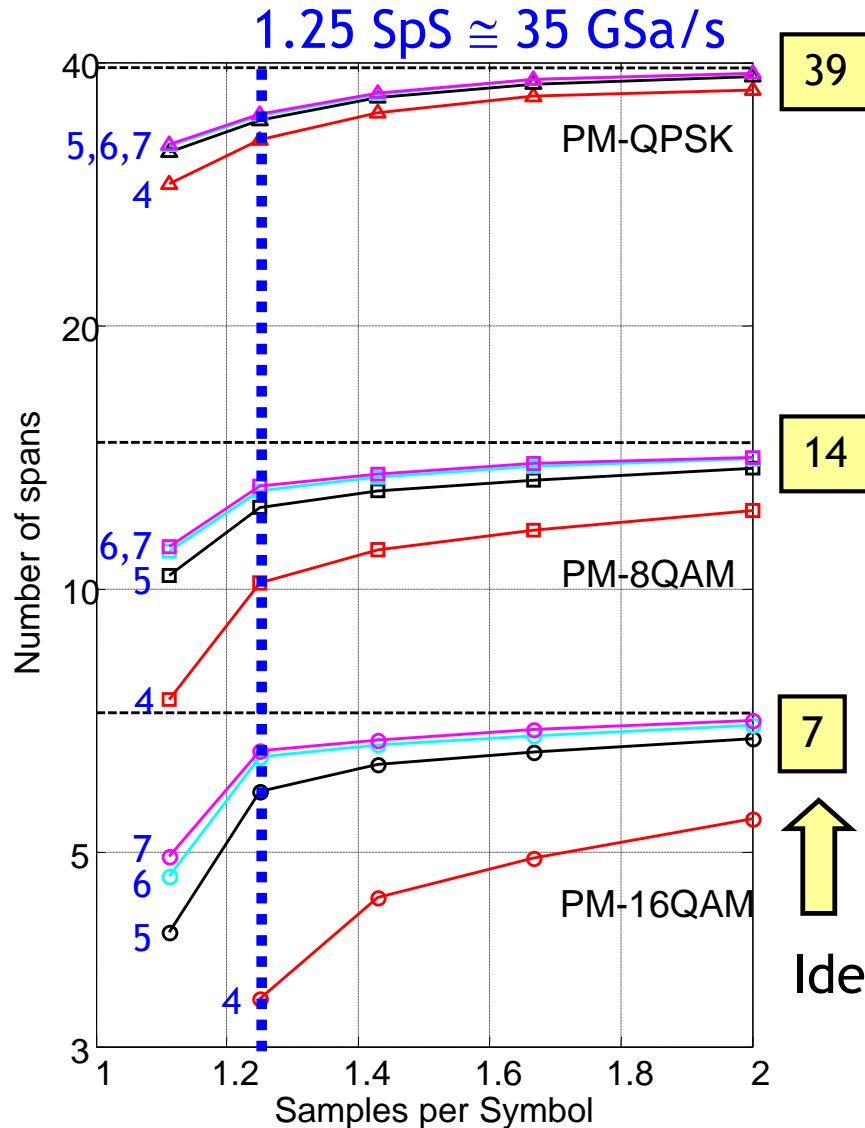
F = 6 dB

- ▶ Uncompensated link
- ▶ Span budget: 25 dB
- ▶ Optimal filter bandwidths
- ▶ We look for maximum reach at BER target, optimizing the launch power

PM-QPSK
SpS=2.00
 $N_{\text{bit}}=7$

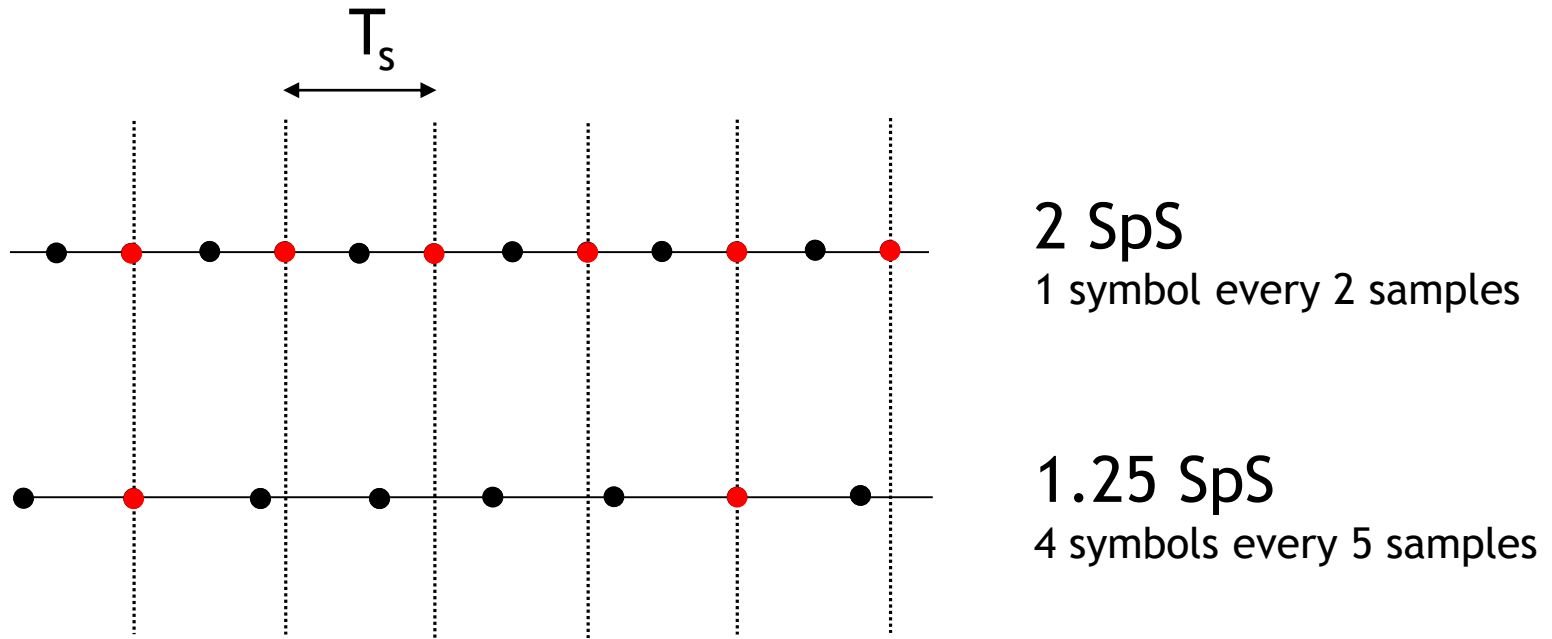


Optimum launch power is not dependent on
ADC speed and resolution



- ▶ Reach reduction is only due to poorer back-to-back performance
- ▶ Assuming to work with $SpS=1.25$ and using $N_{bit}=5$ we get:

	Capacity in C band [Tbps]	Reach	
		Span	Length [km]
PM-QPSK	8	35	3150
PM-8QAM	12	12	1080
PM-16QAM	16	6	540



Equalizer update performed only on **RED** points:

- at 2 SpS EQ is updated every 2 samples
- at 1.25 SpS EQ is updated every 5 samples



DSP running at 1.25 SpS: penalty?



- ▶ Back-to-Back performance
 - ▶ Sampling at 1.25 SpS
 - ▶ with DSP running at 2.00 SpS → 12.25 dB
 - ▶ with DSP running at 1.25 SpS → 12.40 dB
- 0.15 dB penalty
- ▶ Maximum Reach with 7 bit ADC resolution is 35 spans in both cases



3. Mitigation of non-linear effects

- ▶ Coherent receiver can take advantage of DSP also to compensate for non-linear effects
- ▶ Several approaches: all very effective on single channel
- ▶ **Complexity is a major issue...**

PM-QPSK

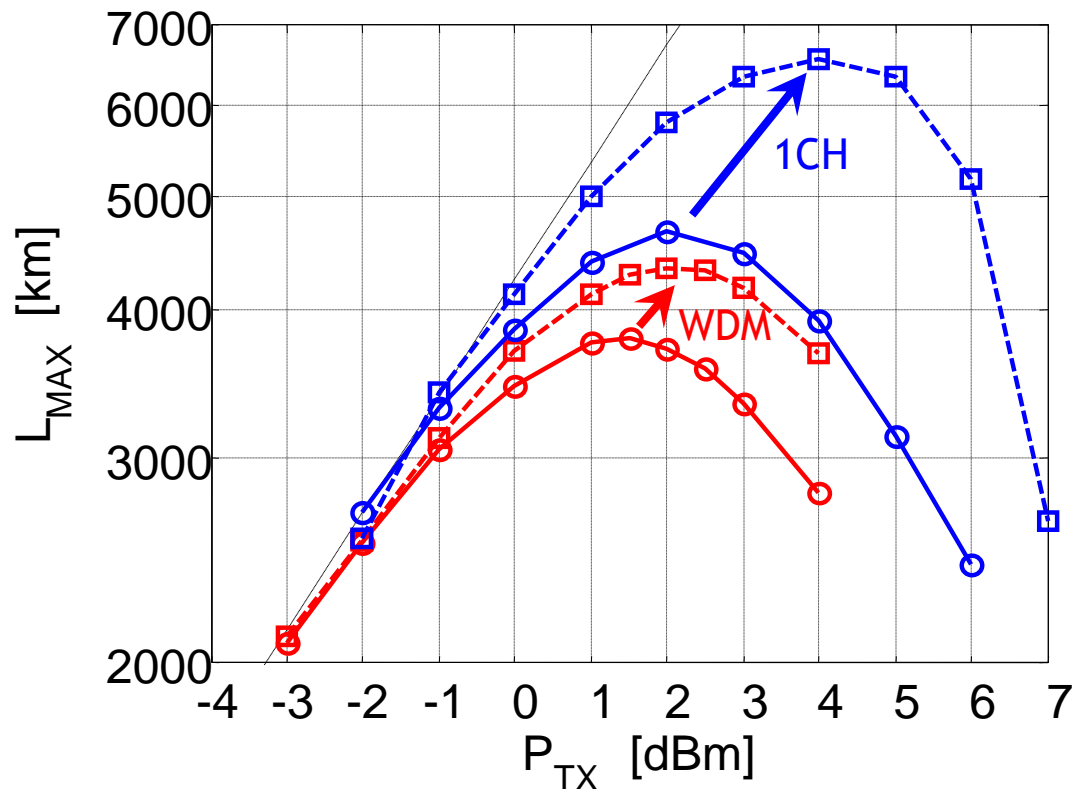
Backward Propagation
(1 step per span)

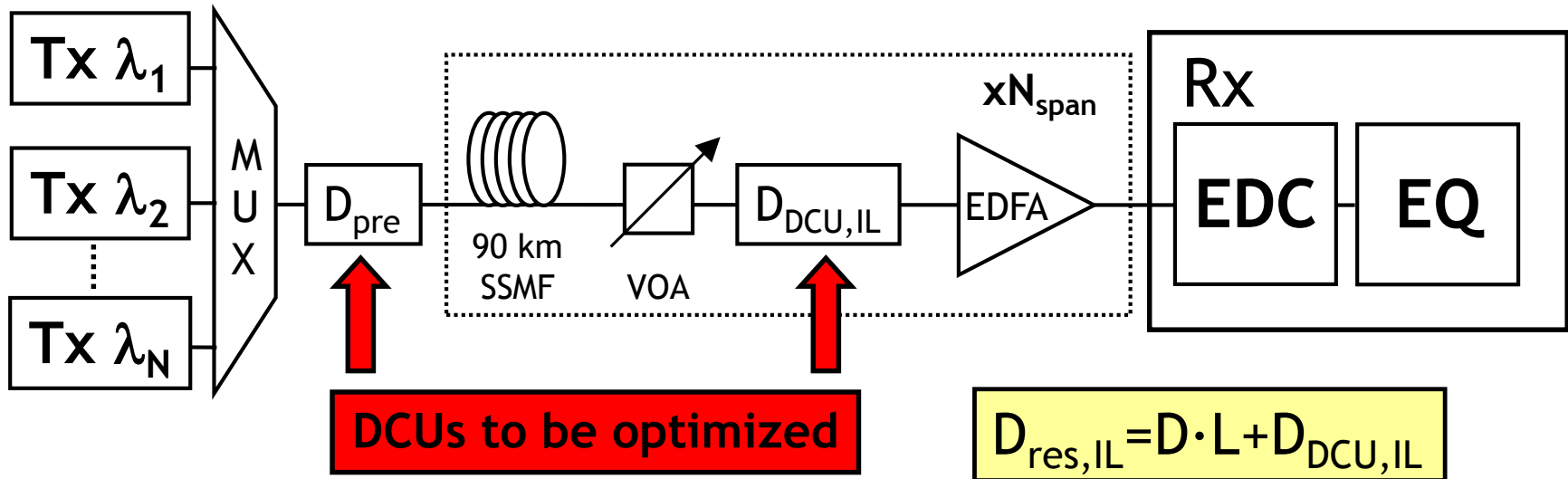
1CH

Improve from 52 to 72 spans → 1.4 dB

WDM

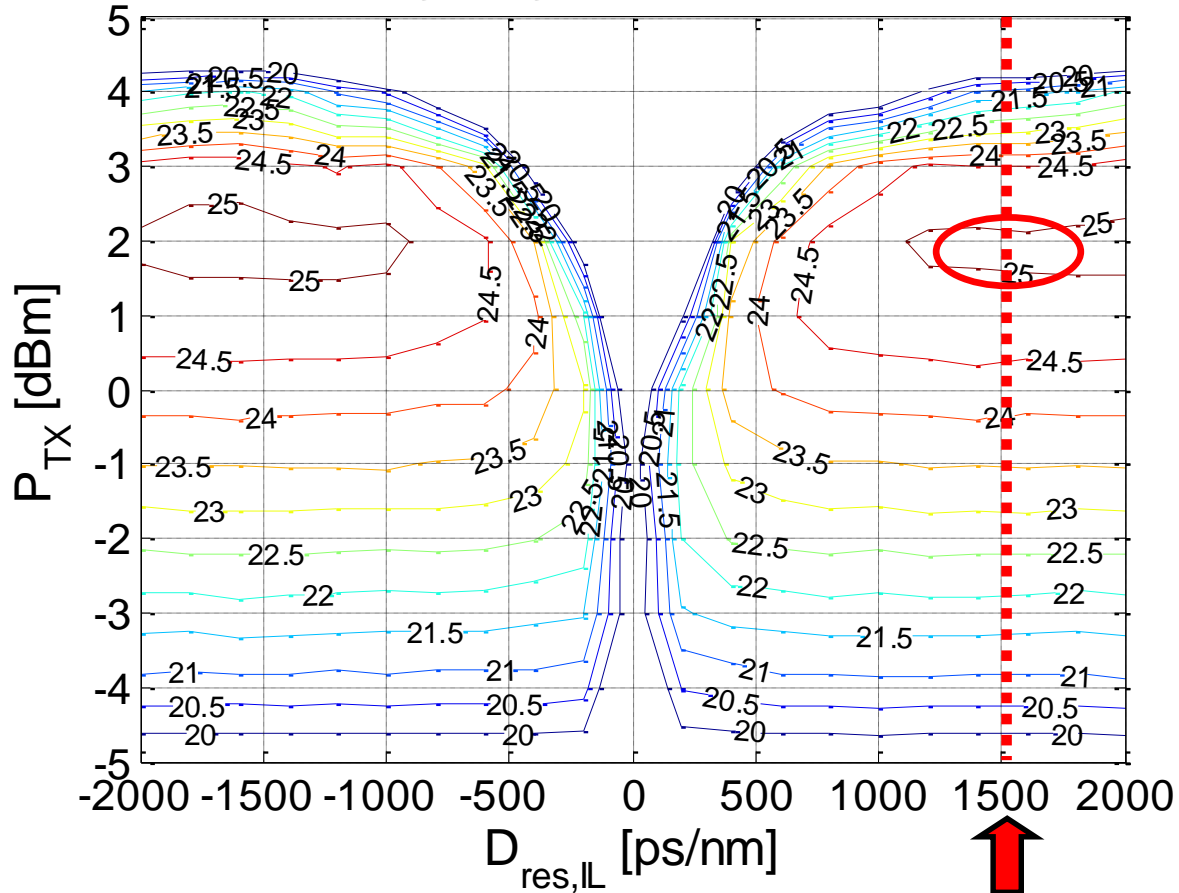
Improve from 39 to 47 spans → 0.8 dB





- ▶ EDC is set to fully compensate total residue
- ▶ Pre-compensation does not give any substantial advantage (see [B]) → $D_{pre} = 0$ ps/nm
- ▶ We carried out joint launch power and in-line residue ($D_{res,IL}$) optimization, looking for maximum Span Budget

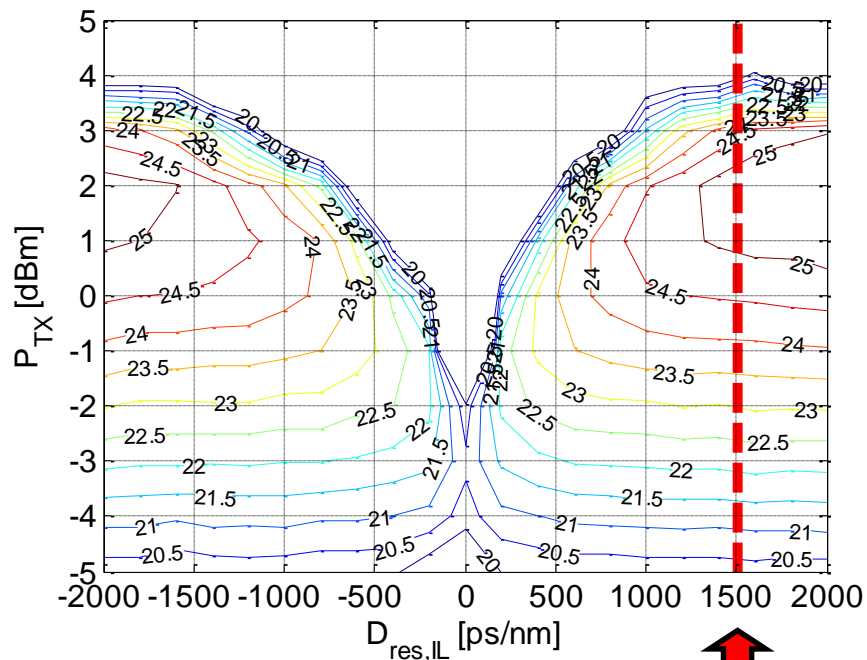
Contour plot of Span Budget
giving $\text{BER}=4 \cdot 10^{-3}$



Without DCU
 $D_{\text{res,IL}} = DL$

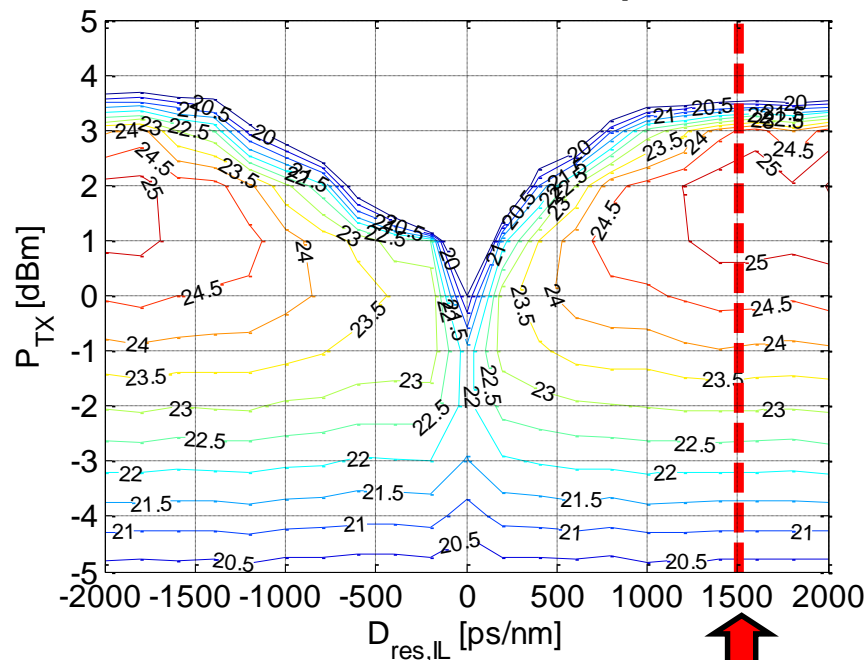
Contour plot of Span Budget

PM-8QAM - $N_{\text{span}}=14$



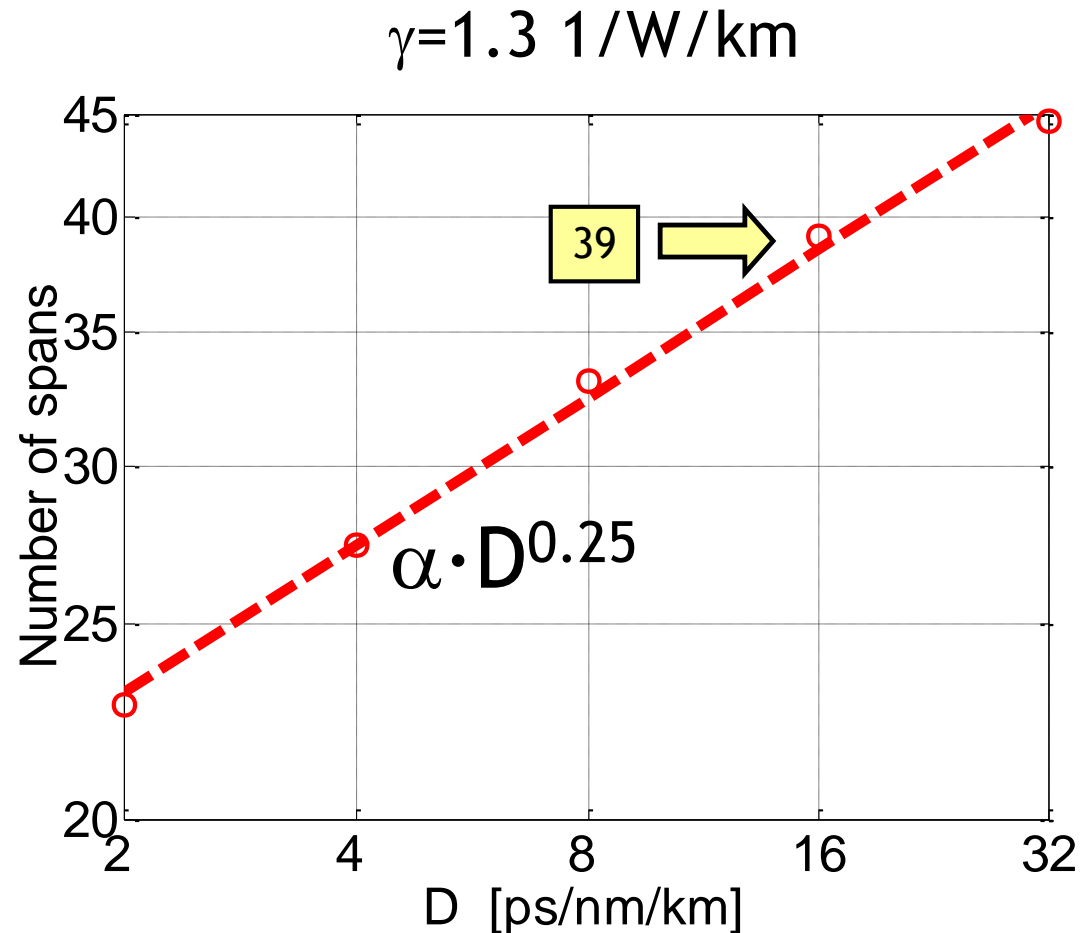
Without DCU
 $D_{\text{res,IL}} = DL$

PM-16QAM - $N_{\text{span}}=7$



Without DCU
 $D_{\text{res,IL}} = DL$

- ▶ New fibers, PSCF in particular, show lower γ , lower α and typically higher dispersion
- ▶ Reducing γ obviously improves the performance
- ▶ Which is the merit of the dispersion itself? For details see [C].



4. Conclusions

▶ RECEIVER REQUIREMENTS

- ▶ ADC speed: using only 1.25 SpS (35 GSa/s) does not cause a substantial penalty
- ▶ ADC resolution: 5 bits are enough
 - ▶ 6 bits is better for PM-16QAM
- ▶ Electrical bandwidth: 10 GHz are enough

▶ NON-LINEAR MITIGATION

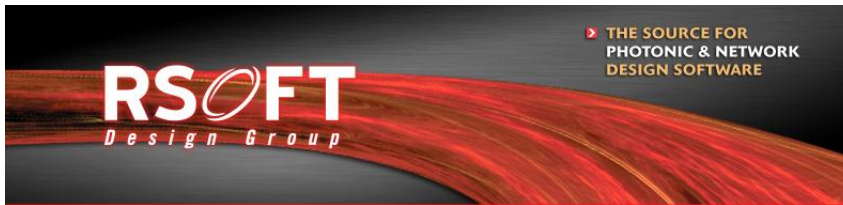
- ▶ Ready
 - ▶ Avoid in-line compensation
 - ▶ New fibers with high dispersion (and low non-linearity)
- ▶ To come
 - ▶ Digital approaches



Acknowledgments



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